US008310801B2

(12) **United States Patent**
McDonald et al.

(10) **Patent No.:** **US 8,310,801 B2**
(45) **Date of Patent:** ***Nov. 13, 2012**

(54) **FLAME SENSING VOLTAGE DEPENDENT ON APPLICATION**

(75) Inventors: **Jonathan McDonald**, Bloomington, MN (US); **Brent Chian**, Plymouth, MN (US)

(73) Assignee: **Honeywell International, Inc.**, Morristown, NJ (US)

(*) Notice: Subject to any disclaimer, the term of this patent is extended or adjusted under 35 U.S.C. 154(b) by 521 days.

This patent is subject to a terminal disclaimer.

(21) Appl. No.: **12/565,676**

(22) Filed: **Sep. 23, 2009**

(65) **Prior Publication Data**

US 2010/0013644 A1   Jan. 21, 2010

Related U.S. Application Data

(63) Continuation-in-part of application No. 10/908,467, filed on May 12, 2005, now Pat. No. 8,066,508, and a continuation-in-part of application No. 12/368,830, filed on Feb. 10, 2009, which is a continuation-in-part of application No. 11/773,198, filed on Jul. 3, 2007, now Pat. No. 8,085,521.

(51) **Int. Cl.**
F23Q 3/00   (2006.01)
(52) **U.S. Cl.** **361/253**
(58) **Field of Classification Search** 361/253
See application file for complete search history.

(56) **References Cited**

U.S. PATENT DOCUMENTS

3,425,780	A	2/1969	Potts
3,520,645	A	7/1970	Cotton et al.
3,649,156	A	3/1972	Conner

3,681,001	A	8/1972	Potts
3,836,857	A	9/1974	Ikegami et al.
3,909,816	A	9/1975	Teeters
4,157,506	A	6/1979	Spencer
4,221,557	A	9/1980	Jalics
4,242,079	A	12/1980	Matthews
4,269,589	A	5/1981	Matthews
4,280,184	A	7/1981	Weiner et al.
4,303,385	A	12/1981	Rudich, Jr. et al.

(Continued)

FOREIGN PATENT DOCUMENTS

EP	0967440	12/1999

(Continued)

OTHER PUBLICATIONS

www.playhookey.com, "Series LC Circuits," 5 pages, printed Jun. 15, 2007.

(Continued)

Primary Examiner — Stephen W Jackson

(74) *Attorney, Agent, or Firm* — Seager, Tufte & Wickhem, LLC

(57) **ABSTRACT**

A system for operating a flame sensing device to obtain readings of increased accuracy without degrading the life of the sensor. There may be levels of a flame requiring a precise measurement. One improvement of accuracy uses higher voltage on the sensor, but this degrades the sensor and thus shortens it life. Further improvement may be achieved by limiting the time that the sensor is operated at a higher voltage. Readings, as if the sensor were operated at a higher voltage, may be inferred from actual readings of the sensor operated at a lower voltage.

20 Claims, 6 Drawing Sheets
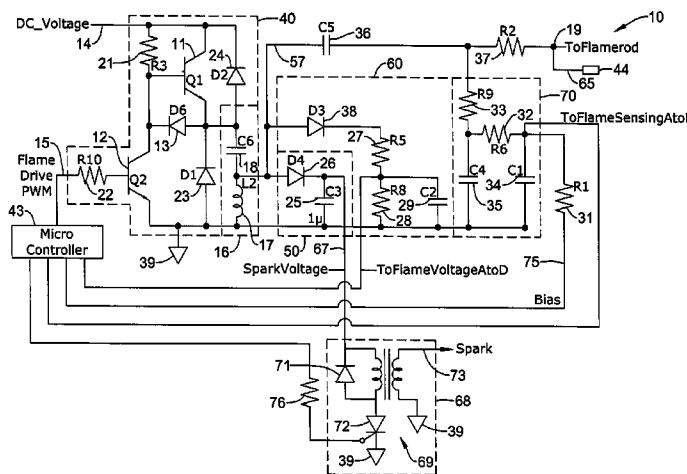
U.S. PATENT DOCUMENTS

4,370,557 A	1/1983	Axmark et al.	
4,450,499 A	5/1984	Sorelle	
4,457,692 A	7/1984	Erdman	
4,483,672 A	11/1984	Wallace et al.	
4,521,825 A	6/1985	Crawford	
4,527,247 A	7/1985	Kaiser et al.	
4,555,800 A	11/1985	Nishikawa et al.	
4,655,705 A	4/1987	Shute et al.	
4,672,324 A	6/1987	van Kampen	
4,695,246 A	9/1987	Beilfuss et al.	
4,709,155 A	11/1987	Yamaguchi et al.	
4,777,607 A	10/1988	Maury et al.	
4,830,601 A	5/1989	Dahlander et al.	
4,842,510 A	6/1989	Grunden et al.	
4,843,084 A	6/1989	Parker et al.	
4,872,828 A	10/1989	Mierzwinski	
4,904,986 A	2/1990	Pinckaers	
4,949,355 A	8/1990	Dyke et al.	
4,955,806 A	9/1990	Grunden et al.	
5,026,270 A	6/1991	Adams et al.	
5,026,272 A	6/1991	Takahashi et al.	
5,037,291 A	8/1991	Clark	
5,073,769 A	12/1991	Kompelien	
5,077,550 A	12/1991	Cormier	
5,112,217 A	5/1992	Ripka et al.	
5,126,721 A	6/1992	Butcher et al.	
5,158,447 A	10/1992	Geary	
5,175,439 A	12/1992	Harer et al.	
5,222,888 A	6/1993	Jones et al.	
5,236,328 A	8/1993	Tate et al.	
5,255,179 A	10/1993	Zekan et al.	
5,276,630 A	1/1994	Baldwin et al.	
5,280,802 A	1/1994	Comuzie, Jr.	
5,300,836 A	4/1994	Cha	
5,347,982 A	9/1994	Binzer et al.	
5,365,223 A	11/1994	Sigafus	
5,391,074 A	2/1995	Meeker	
5,424,554 A	6/1995	Marran et al.	
5,446,677 A	8/1995	Jensen et al.	
5,472,336 A	12/1995	Adams et al.	
5,506,569 A	4/1996	Rowlette	
5,567,143 A	10/1996	Servidio	
5,599,180 A	2/1997	Peters et al.	
5,682,329 A	10/1997	Seem et al.	
5,722,823 A	3/1998	Hodgkiss	
5,797,358 A	8/1998	Brandt et al.	
5,971,745 A	10/1999	Bassett et al.	
6,060,719 A	5/2000	DiTucci et al.	
6,071,114 A	6/2000	Cusack et al.	
6,084,518 A	7/2000	Jamieson	
6,222,719 B1	4/2001	Kadah	
6,261,086 B1	7/2001	Fu	
6,299,433 B1	10/2001	Gauba et al.	
6,346,712 B1	2/2002	Popovic et al.	
6,349,156 B1	2/2002	O'Brien et al.	
6,356,827 B1	3/2002	Davis et al.	
6,385,510 B1	5/2002	Hoog et al.	
6,457,692 B1	10/2002	Gohl, Jr.	
6,474,979 B1	11/2002	Rippelmeyer	
6,486,486 B1	11/2002	Haupenthal	
6,509,838 B1	1/2003	Payne et al.	
6,552,865 B2	4/2003	Cyrusian	
6,676,404 B2	1/2004	Lochschmied	
6,743,010 B2	6/2004	Bridgeman et al.	
6,782,345 B1	8/2004	Siegel et al.	
6,794,771 B2	9/2004	Orloff	
6,912,671 B2	6/2005	Christensen et al.	
6,917,888 B2	7/2005	Logvinov et al.	
7,088,137 B2	8/2006	Behrendt et al.	
7,088,253 B2	8/2006	Grow	
7,202,794 B2	4/2007	Huseynov et al.	
7,241,135 B2	7/2007	Munsterhuis et al.	
7,255,285 B2	8/2007	Troost et al.	
7,274,973 B2	9/2007	Nichols et al.	
7,289,032 B2	10/2007	Seguin et al.	
7,327,269 B2	2/2008	Kiarostami	
7,617,691 B2	11/2009	Street et al.	
7,728,736 B2	6/2010	Leeland et al.	
7,764,182 B2	7/2010	Chian et al.	
7,768,410 B2	8/2010	Chian	
7,800,508 B2 *	9/2010	Chian et al.	340/577
8,066,508 B2 *	11/2011	Nordberg et al.	431/75
8,085,521 B2 *	12/2011	Chian	361/253
2002/0099474 A1	7/2002	Khesin	
2003/0064335 A1	4/2003	Canon	
2003/0222982 A1	12/2003	Hamdan et al.	
2004/0209209 A1	10/2004	Chodacki et al.	
2005/0086341 A1	4/2005	Enga et al.	
2006/0257805 A1	11/2006	Nordberg et al.	
2007/0159978 A1	7/2007	Anglin et al.	
2007/0188971 A1	8/2007	Chian et al.	
2009/0009344 A1	1/2009	Chian	
2009/0136883 A1	5/2009	Chian et al.	
2010/0013644 A1	1/2010	Mcdonald et al.	
2010/0265075 A1	10/2010	Chian	

FOREIGN PATENT DOCUMENTS

EP	1148298	10/2001
WO	9718417	5/1997

OTHER PUBLICATIONS

Honeywell, "SV9410/SV9420; SV9510/SV9520; SV9610/SV9620 SmartValve System Controls," Installation Instructions, 16 pages, 2003.

Honeywell, "S4965 SERIES Combined Valve and Boiler Control Systems," 16 pages, prior to the filing date of present application.

Honeywell, "SV94101SV9420; SV9510/SV9520; SV9610/SV9620 SmartValve System Controls," Installation Instructions, 16 pages, 2003.
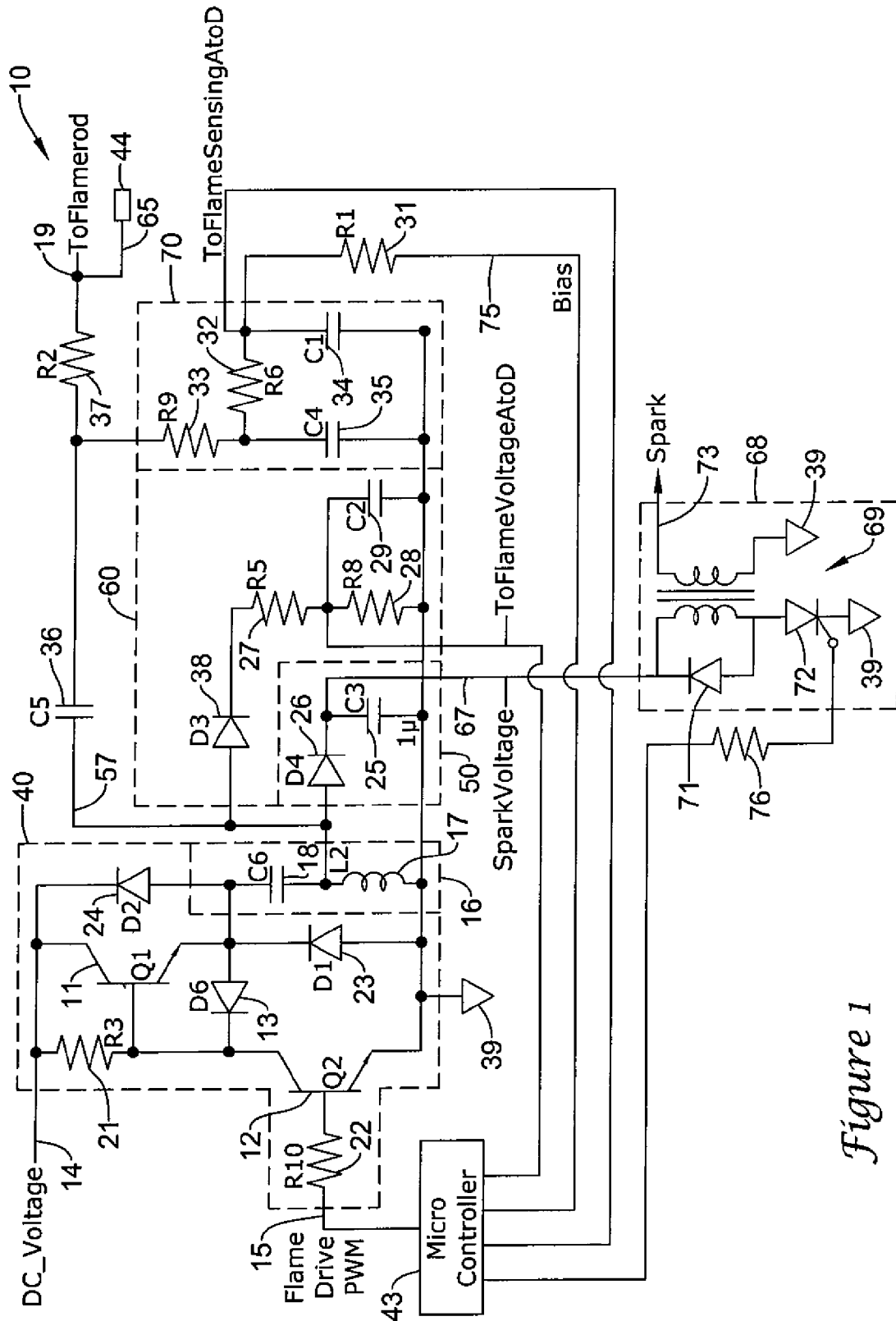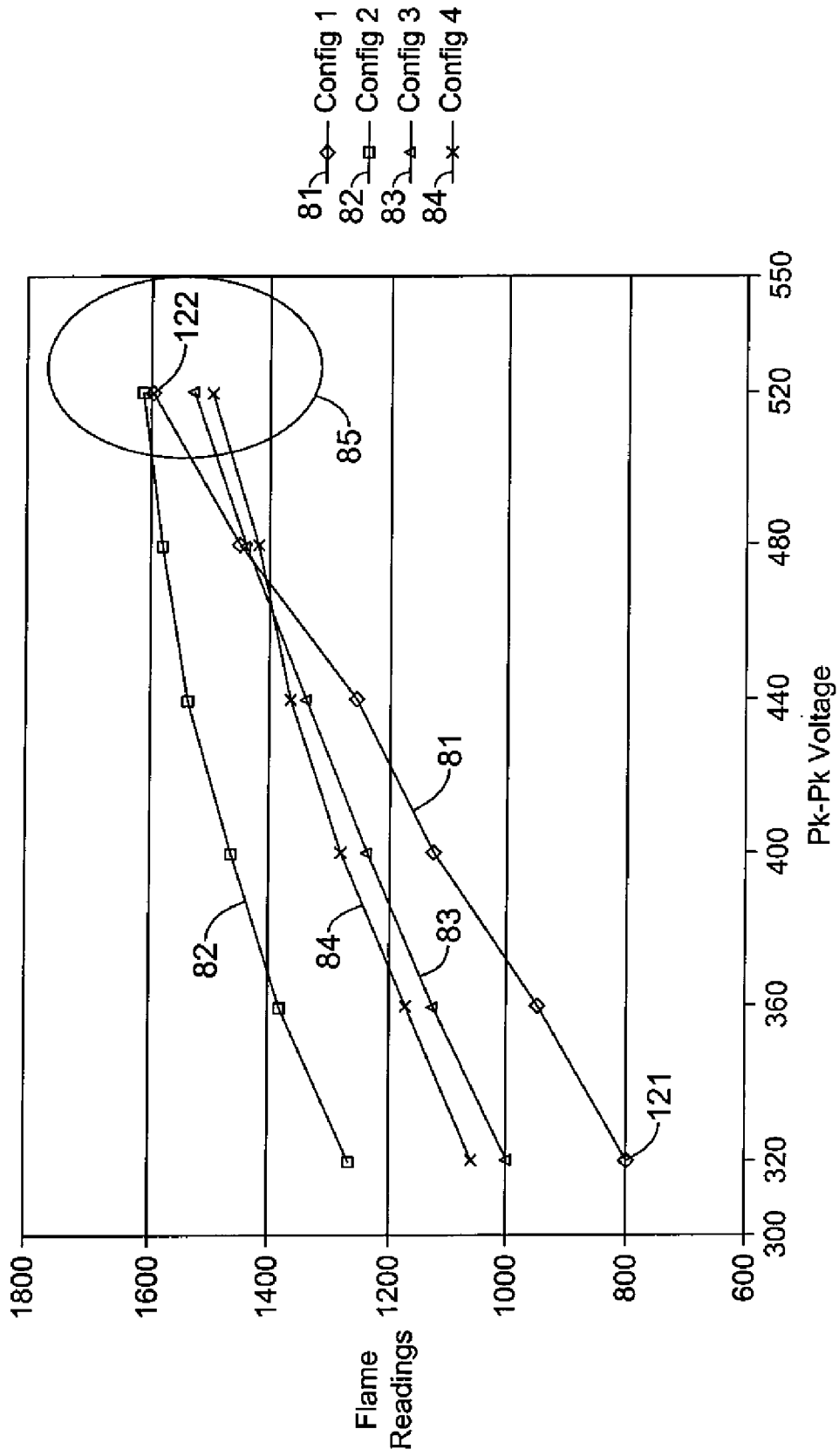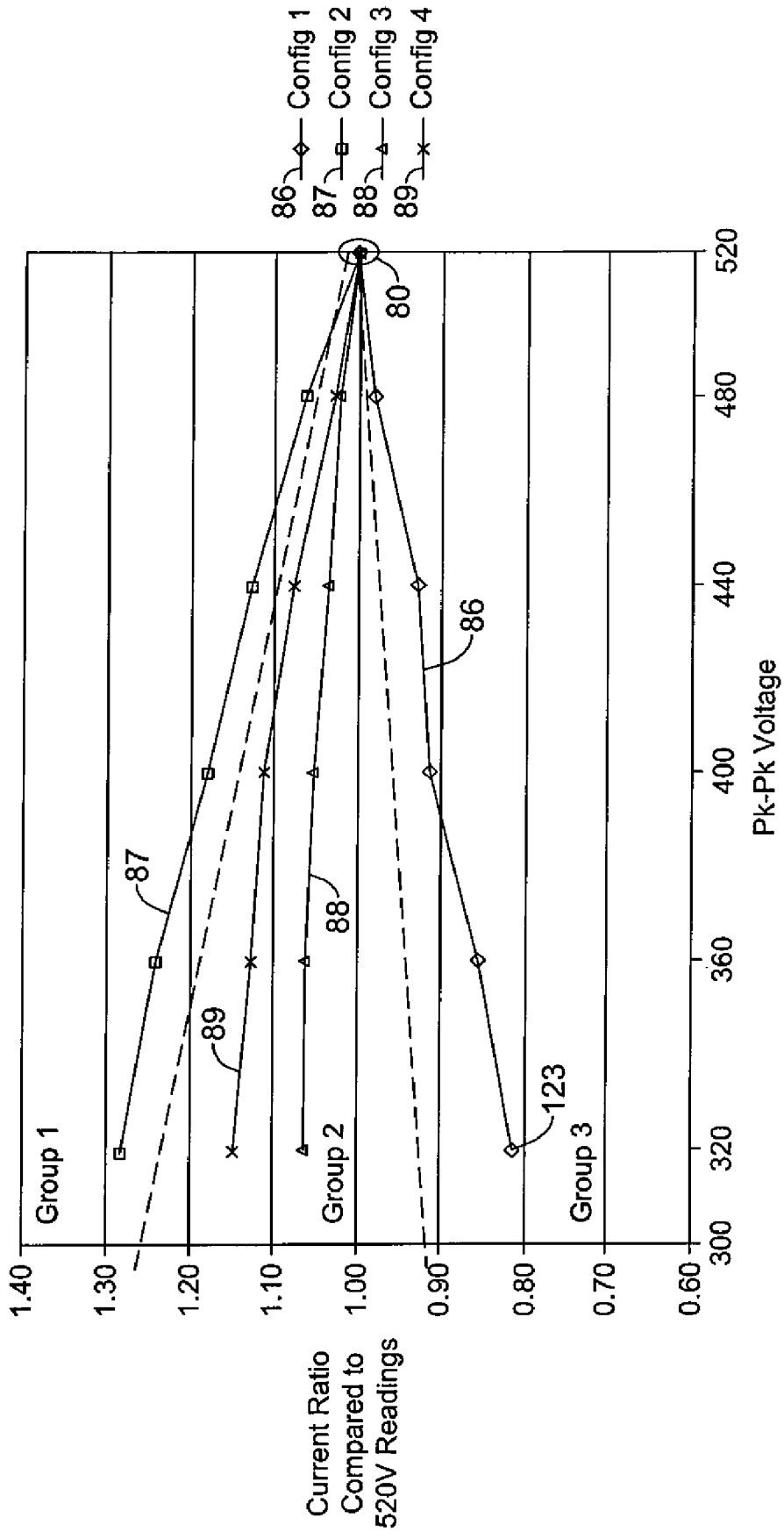
* cited by examiner

Figure 1

Figure 2

Figure 3

90

91

Flame in critical range?

yes

92

V_flame = High Voltage

93

Flame_scaling = High

no

94

Flame > High Flame Threshold?

yes

95

V_flame = Low Voltage

96

Flame_scaling = Low

no

97

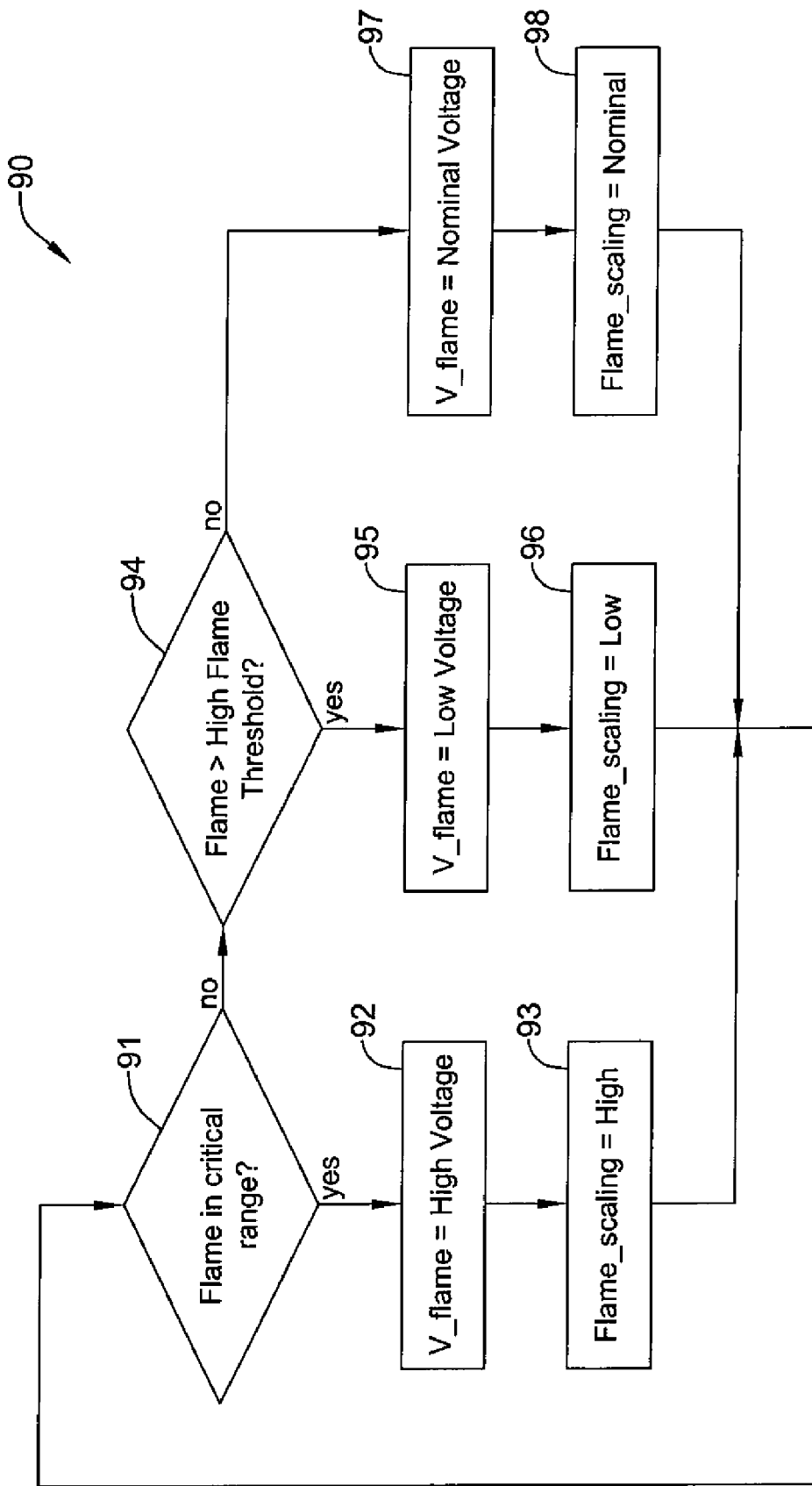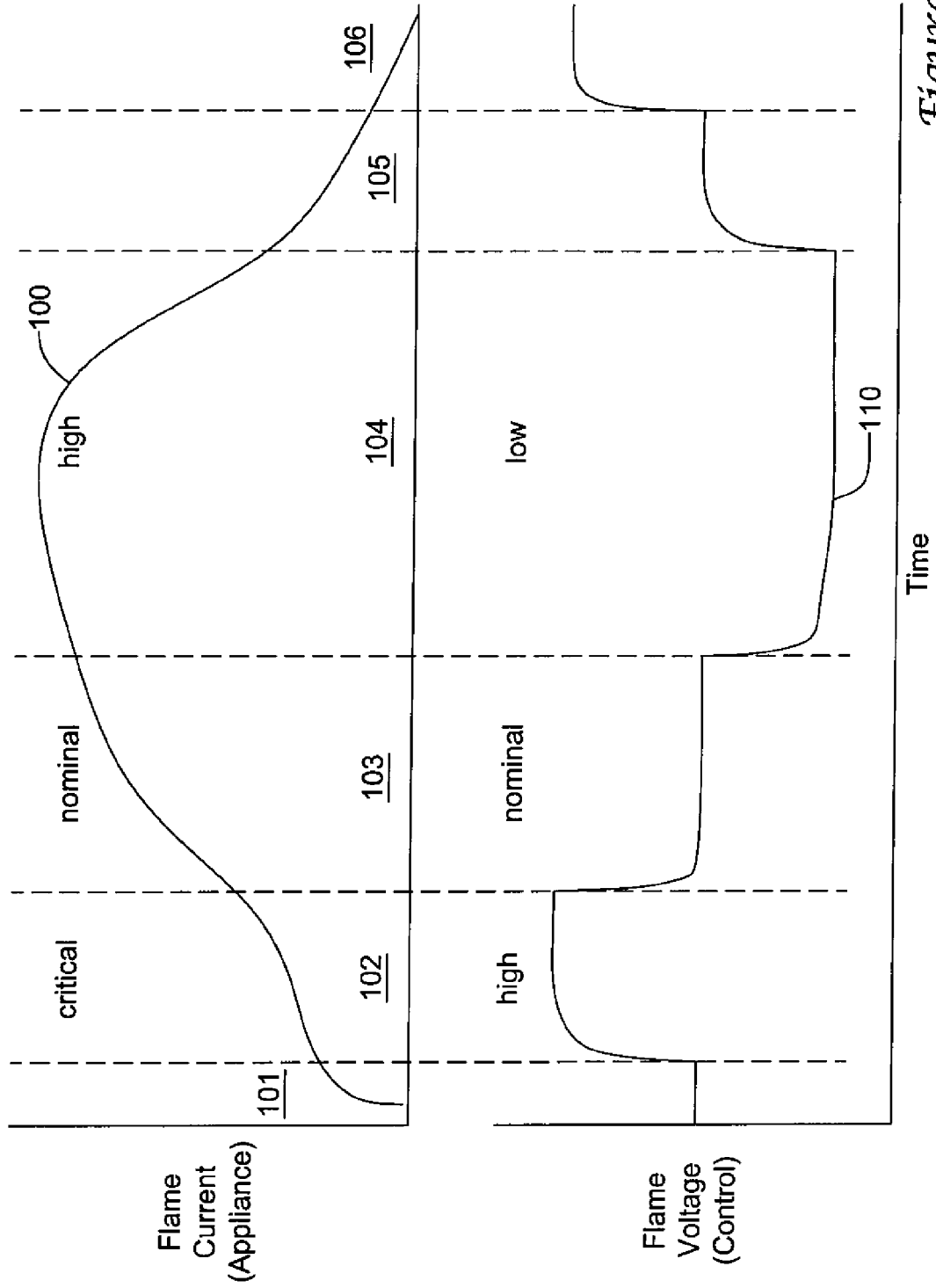V_flame = Nominal Voltage

98
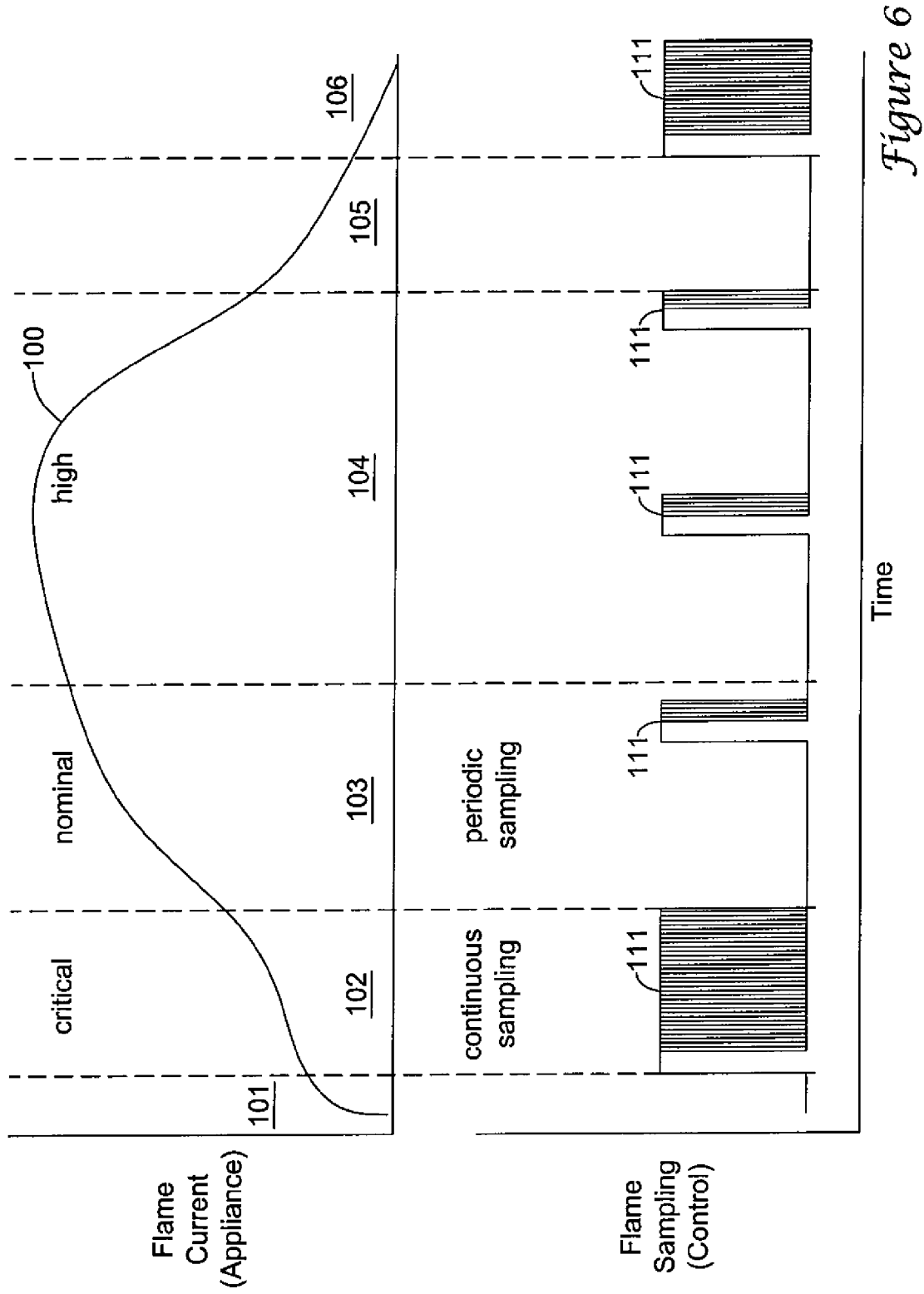
Flame_scaling = Nominal

Figure 4

Figure 5

Figure 6

FLAME SENSING VOLTAGE DEPENDENT ON APPLICATION

The present application is a continuation-in-part of U.S. patent application Ser. No. 10/908,467, filed May 12, 2005, and entitled "Adaptive Spark Ignition and Flame Sensing Signal Generation System". U.S. patent application Ser. No. 10/908,467, filed May 12, 2005, and entitled "Adaptive Spark Ignition and Flame Sensing Signal Generation System", is hereby incorporated by reference.

The present application is a continuation-in-part of U.S. patent application Ser. No. 12/368,830, filed Feb. 10, 2009, and entitled "Low Cost High Speed Spark Voltage and Flame Drive Signal Generator", which in turn is a continuation-in-part of U.S. patent application Ser. No. 11/773,198, filed Jul. 3, 2007, and entitled "Flame Rod Drive Signal Generator and System". U.S. patent application Ser. No. 12/368,830, filed Feb. 10, 2009, and entitled "Low Cost High Speed Spark Voltage and Flame Drive Signal Generator", is hereby incorporated by reference. U.S. patent application Ser. No. 11/773, 198, filed Jul. 3, 2007, and entitled "Flame Rod Drive Signal Generator and System", is hereby incorporated by reference.

RELATED APPLICATIONS

The present application is related to the following indicated patent applications: U.S. patent application Ser. No. 11/741, 435, filed Apr. 27, 2007, and entitled "Combustion Instability Detection"; U.S. patent application Ser. No. 11/276,129, filed Feb. 15, 2006, and entitled "Circuit Diagnostics from Flame Sensing AC Component"; U.S. patent application Ser. No. 11/306,758, filed Jan. 10, 2006, and entitled "Remote Communications Diagnostics Using Analog Data Analysis"; U.S. patent application Ser. No. 10/908,466, filed May 12, 2005, and entitled "Flame Sensing System"; U.S. patent application Ser. No. 10/908,465, filed May 12, 2005, and entitled "Leakage Detection and Compensation System"; U.S. patent application Ser. No. 10/908,463, filed May 12, 2005, and entitled "Dynamic DC Biasing and Leakage Compensation"; and U.S. patent application Ser. No. 10/698,882, filed Oct. 31, 2003, and entitled "Blocked Flue Detection Methods and Systems"; all of which are incorporated herein by reference.

BACKGROUND

The invention pertains to sensors and particularly to flame sensors. More particularly, the invention pertains to optimization of flame sensing.

SUMMARY

The invention is a system for operating a flame sensing device to obtain readings of increased accuracy without degradation of the life of the sensor.

BRIEF DESCRIPTION OF THE DRAWING

FIG. 1 is a diagram of a spark voltage and flame signal generation circuit;

FIG. 2 is a graph showing flame current from four different flame rod configurations over a wide voltage range;

FIG. 3 is a graph showing an approach for improved accuracy of flame sensing without a need for continuous high voltage;

FIG. 4 is a flow diagram of a control system for flame sensing;

FIG. 5 is a graphic example of the voltage adjustment of the control system described in FIG. 4 based on a typical appliance run cycle; and

FIG. 6 is a graphic example of the control sampling of the flame signal at various times or zones during an appliance run cycle.

DESCRIPTION

The flame current sensed in an ignition system may depend on the applied voltage. In particular, the relationship between AC voltage and flame current at a given frequency may be different for each application. Not only does this result in less accurate flame readings, but could create a safety concern if not handled properly. In addition, using too high of an AC voltage may cause excessive build-up of contamination on a flame rod, increased energy consumption that generates extra heat, and also stress associated electronic circuitry unnecessarily.

One possibility for more accurately measuring the flame signal at a given frequency may be to increase the AC voltage when accuracy is critical. It appears that higher voltages reduce the overall differences between different flame rod configurations. Once a flame has been established, the AC voltage may be adjusted to a lower level to avoid excessive component stress, energy consumption, increased electrical noise, and contamination build-up.

Another approach may be to vary the AC voltage in order to generate a curve of flame readings for a particular flame rod configuration. Once this curve or ratio between different voltages has been determined at a given flame level, a lower AC voltage may be used and the flame sensed value can be scaled as needed.

An electronic circuit with adjustable AC voltage supply may be used to generate the different voltage levels. This may be accomplished using a resonant circuit such as an inductor-capacitor combination driven at varying duty cycles with a feedback network used to fine-tune the voltage level. The software in an embedded microprocessor may then adjust the AC voltage to the highest level required, say 250 Vpk, for most accurate flame sensing, and can readjust to a lower level, say 170 Vpk or 90 Vpk, to sense less critical flame levels and help extend the life of the system. Other voltage levels may be used, depending on the particular flame sensing apparatus.

Alternatively, the microprocessor may switch between different voltage levels very quickly and compare the flame readings at each level to determine a ratio factor. Using this ratio factor, the measured flame current at lower voltage levels may be scaled to an equivalent higher voltage reading or via a predetermined lookup table, based on empirical or calculated data, for greater accuracy.

Either method may limit the amount of time using the highest voltage levels, thus reducing component stress and noise, limiting energy consumption, and improving life of the flame rod with reduced contamination build-up.

FIG. 1 is a diagram of a spark and flame signal generation circuit 10. Transistors 11 and 12 and diode 13 form a push-pull drive. DC_voltage 14 relative to a reference terminal or ground 39 may be rectified 24VAC. Voltage 14 may be in the range of 20 to 40 volts. When FlameDrivePWM 15 is at a resonant frequency of the LC circuit 16 containing an inductor 17 and capacitor 18, a high voltage near sinusoidal waveform may be generated as an output 57 at the common node of inductor 17 and capacitor 18. The common node or output of circuit 16 may be also regarded as an output terminal 57. Inductor 17 may have value of about 18 millihenries and capacitor 18 may have a value of about 10 nanofarads. A duty

cycle of FlameDrivePWM **15** may be changed with pulse width modulation to control the amplitude of the near sinusoidal waveform. The waveform may be sent to ToFlameRod terminal **19** connected via a D.C. blocking capacitor **36** and current limiting resistor **37**. The waveform may proceed from terminal **19** via a line **65** to a flame rod **44** for flame sensing. Capacitor **36** may have a value of about 2,200 picofarads. Resistor **37** may have a value of about 100 K-ohms.

A high level voltage does not necessarily exist anywhere in the drive circuit **40** (a 1.5 K-ohm resistor **21**, a 2 K-ohm resistor **22**, diode **23**, diode **24**, diode **13**, transistor **11** and transistor **12**). So these components may be implemented for low voltage applications and have a low cost.

Diode **23** and diode **24** may be added to provide current path when the resonant current of the LC network **16** is not in perfect synchronization with the drive signal. To generate a spark voltage on capacitor **25** quickly, the drive may need to be rather strong, and diode **23** and diode **24** may be added to improve the network efficiency and reduce the heat generated on the drive components.

A spark voltage circuit **50** may include components **25** and **26**. Diode **26** may rectify the AC output voltage from circuit **16** so as to charge up a capacitor **25**. Capacitor **25** may be charged up to a high voltage level for spark generation. Typically, capacitor **25** may be 1 microfarad and be charged up to about 170 volts or so for each spark.

An output **67** of circuit **50** may go to a spark circuit **68**. Output **67** may be connected to a first end of a primary winding of a transformer **69** and to a cathode of a diode **71**. An anode of diode **71** may be connected to a second end of the primary winding. The second end of the primary winding may be connected to an anode of an SCR **72**. A cathode of SCR **72** may be connected to a reference voltage or ground **39**. A gate of SCR **72** may be connected to controller **43** through a 1 K-ohm resistor **76**. A first end of a secondary winding of transformer **69** may be connected to a spark terminal **73**. A second end of the secondary winding of transformer **69** may be connected to ground or reference voltage **39**.

When capacitor **25** is charged up, a signal from controller **43** may go to the gate of SCR **72** to turn on the SCR and discharge capacitor **25** to ground or reference voltage **39** resulting in a high surge of current through the primary winding of transformer **69** to cause a high voltage to be across the secondary winding to provide a spark between terminal **73** and ground or reference voltage **39**.

A diode **38**, a 470 K-ohm resistor **27**, a 35.7 K-ohm resistor **28** and a 0.1 microfarad capacitor **29** may form a circuit **60** for sensing flame voltage from output **57** of LC circuit **16**. Circuit **60** may provide an output signal, from the common connection of resistors **27** and **28** to microcontroller **43**, indicating the voltage amplitude of the drive signal to flame rod **44**.

A 200 K-ohm resistor **32**, a 200 K-ohm resistor **33**, a 0.01 microfarad capacitor **34** and a 0.01 microfarad capacitor **35** may form a circuit **70** having an output at the common connection of resistor **32** and capacitor **34** for flame sensing which goes to controller **43**. At least a portion of circuit **70** may incorporate a ripple filter for filtering out the AC component of the flame rod drive signal so as to expose the DC offset current of flame rod **44**. The DC offset current may be indicated at the output of circuit **70**. When a flame is present, flame rod **44** may have a corresponding DC offset current. A resistor connected in series with a diode having its cathode connected to ground may be an equivalent circuit of flame rod **44** sensing a flame. When no flame is present, flame rod **44** may have no or little DC offset current. Resistor **31** may be a bias element. Microcontroller **43** may provide a bias **75** input

(e.g., about 4.5 volts) to circuit **70** via a 200 K-ohm resistor **31**. As the flame current is flowing from flame rod **44** out to the flame, generating a negative voltage at capacitor **34**, a positive bias **75** is necessary to pull the voltage at capacitor **34** above ground or reference voltage **39** for microcontroller **43** to measure the flame.

At first power up, a microcontroller **43** may drive a FlameDrivePWM signal at an input **15** with a nearly square waveform shape. The frequency of the FlameDrivePWM signal at terminal **15** may be varied and the flame voltage at line **57** be monitored to find the resonant frequency of the LC network **16**. After that, the drive is generally kept at this frequency, and the duty cycle may be changed so that capacitor **25** can be charged to the required level within the predetermined time interval. This duty cycle may be stored as SparkDuty. The duty cycle may be changed again to find a duty cycle value at which the flame sensing signal is at the desired level, for example, 180 volts peak. This duty cycle value may be saved as FlameDuty. The frequency of the PWM signal **15** may be changed to fine tune the signal amplitude at the output of LC network **16**.

One may note that if the DC_Voltage **14** changes, the duties may need adjustment. This adjustment may be done continuously and slowly at run time. At spark time, the FlameDrivePWM signal may stay at the SparkDuty value and the spark voltage be monitored. The SparkDuty value may be adjusted as necessary during spark time.

At flame sensing time, capacitor **25** is to be overcharged some 10 to 20 volts higher than the flame voltage, so that capacitor **25** will not present itself as a burden or heavy load on the LC network **16** and thus the flame voltage at line **57** can be varied quickly.

The flame sensing circuit **70** may support a high flame sensing rate, such as 60 samples per second. Sixty samples/ second may be limited by the fact that the drive and flame signal itself carries a line frequency component, not limited by the circuit.

FIG. **2** is a graph showing an example of typical flame readings (taken at one flame level) from four different flame rod configurations over a wide voltage range. Data may be empirically obtained by taking flame readings at various voltages for each of the several configurations, and plotted on a graph like that in FIG. **2** or recorded and arranged in another manner. The flame readings versus peak-to-peak (Pk-Pk) voltage for configurations **1**, **2**, **3** and **4** are plotted as revealed by curves **81**, **82**, **83** and **84**, respectively. A high voltage flame circuit as described in FIG. **1** may be used to generate the high voltage needed for flame rectification. As the graph shows, expected accuracy at a flame excitation voltage of 320V pk-pk is about +/−20 percent. At 520V pk-pk, the accuracy improves to better than +/−5 percent at area **85**. Whenever accuracy of the flame readings is critical, the highest excitation voltage could be used. When flame readings are high and accuracy is less critical, lower excitation voltages may be used to reduce power consumption and noise, extend life of electrical components, and reduce contamination build-up on the flame rod **44**.

FIG. **3** is a graph showing an approach to gain improved accuracy without the need for continuous flame sensing at a high excitation voltage. The approach includes measuring the flame at a lower voltage and scaling the flame readings to an equivalent higher voltage flame level. A current ratio to 520V readings versus lower Pk-Pk voltages at a given flame level is graphed in FIG. **3** for four different flame rod configurations. To determine which scaling factor to use, a comparison of the flame readings at two different voltages may be done resulting in a "current ratio." For example, in this graph, configu-

ration 1 has a current ratio between 320V pk-pk and 520V pk-pk of just over 0.80, as shown by curve **86**, while configuration 2 has a ratio of just less than 1.30, as shown by curve **87**. The ratios for configurations **3** and **4** are shown by curves **88** and **89**. Data in the graph of FIG. **2** may be used to determine the ratios plotted in the graph of FIG. **3**. These current ratios may be used to directly scale a lower voltage flame reading to their equivalent higher voltage levels. Another implementation of this scaling may include dividing the current ratios into predetermined groups **1** through **3**, as shown in FIG. **3**. Group **2** may include both configurations **3** and **4**, represented by curves **88** and **89**, respectively, since their current ratios are very close, and as expected in FIG. **2** their actual flame readings are very close. Group **1** may include curve **87** and group **3** may include curve **86**. Additional data may be taken and other calculations made for plotting points on the graphs in FIGS. **2** and **3** for different flame rod configurations. Since the ratios in FIG. **3** are based on 520 volts pk-pk readings, the ratios of the configurations converge to one at that level as indicated at area **80**. Additional current levels other than those shown in FIGS. **2** and **3** may be used for calculating the flame scaling ratios. These measurements can be referenced by any equivalent voltage units as appropriate, such as pk-pk, pk or rms. Since the ratios shown are for one particular flame level, additional ratios may be calculated to cover the entire operating range of flame currents for greatest accuracy.

The approach for using low voltages to obtain high voltage-like readings may require an initial calibration period when the voltage levels are quickly changed between high and low levels; but once the respective current ratio is established, control may be allowed to run at a low excitation voltage and result in reduced stress on components as noted herein.

A formula may be used for various calculations related to flame sensing. R_{H1} may be regarded as a relatively accurate flame reading of a flame sensor, for example, configuration **1** at a designated high voltage. V_H may represent the designated high voltage for the sensor at a flame reading in the area **85** of FIG. **2**, which may be regarded as a relatively accurate area of flame readings from flame sensors of various configurations. R_{L1} may be a flame reading of a flame sensor of the configuration **1** taken at a sensor voltage V_L which would have a magnitude less than that of V_H. A flame reading divided by the sensor voltage may be a ratio. For example, r_{L1} may represent the ratio for R_{L1}/V_L and r_{H1} may represent the ratio for R_{H1}/V_H involving a flame sensor of configuration **1**. A current ratio relative to the V_H flame reading for configuration **1** may be designated as r_{C1} which may equal r_{L1}/r_{H1} or $(R_{L1}/V_L)/(R_{H1}/V_H)$.

For instance, to calculate the reading-to-voltage ratio (r_{L1}) for configuration **1** at a reading for a pk-pk voltage of 320 (V_L), one may note a flame reading of 800 units (R_{L1}), as shown by point **121** on curve **81** in FIG. **2**. A reading-to-voltage ratio (r_{H1}), and for a pk-pk voltage of 520 (V_H), one may note a reading of about 1600 units (R_{H1}) at point **122** on curve **81**. One may divide 800 units by 320 volts to obtain 2.50 units per volt (r_{L1}), and divide 1600 units by 520 volts to obtain about 3.08 units per volt (r_{H1}). To obtain the current ratio for the readings of configuration **1** at 320 volts and 520 volts, one may divide the 2.50 flame reading units per volt at the 320 volt reading by the 3.08 flame reading units per volt at the 520 volt reading to obtain a current ratio of about 0.8125 (r_{C1}). This ratio may be plotted as point **123** as part of plot or curve **86** for configuration **1** on the graph in FIG. **3**. The flame reading at 520 volts may be regarded as the most precise reading (e.g., a touchstone) since the readings of all the configurations may converge at area **85**. With the current ratio (r_{C1}) for a flame reading from a flame sensor of configuration

1 at a low 320 volt level, one may calculate, scale or extrapolate a relatively precise flame reading at a high 520 volt level. One may take the r_{C1} equation and derive $R_{H1}=(R_{L1}V_H)/(r_{C1}V_L)$. If a low voltage reading (V_L) is 800; calculating for the reading R_{H1} as it should be with the high sensor voltage V_H, one may get $(800\times520)/0.8125\times320)=1600$. One may convert other readings at the low voltage for obtaining readings as they would be if obtained at the high voltage. The present approach may be used for obtaining readings for other configurations and voltages. This portion of the approach may be in a look-up table, program, or other form of control. The general approach may be in a look-up table, program, input, or other form of stored control or processing. An advantage of the approach is that without actually running a flame rod and associated components at the high voltage, one may still obtain high-voltage precision readings and avoid excessive component stress, energy consumption and contamination build-up which would occur when obtaining flame readings using high voltage on the flame sensor.

Similar calculations for current ratios may be done for other flame readings at other voltages for the flame sensor or sensing rod **44** (FIG. **1**) of configuration **1**. Flame readings may be taken for configurations **2**, **3** and **4** as shown in the graph of FIG. **2**. Calculations may be performed to obtain current ratios for flame sensor or sensing rod configurations **2**, **3** and **4**, and be plotted as shown in the graph of FIG. **3**. Data and calculations may be obtained and plotted for other configurations. The voltages used may also be different. In summary, the information of FIGS. **2** and **3** may be used for obtaining flame readings measured at lower voltages which are nearly as accurate as if these readings were measured at optimally higher voltages. FIGS. **2** and **3** were plotted for one flame level (i.e., 0.7 micro amp). At other flame current levels, the curves may be different. Thus, FIGS. **2** and **3** may be plotted for other flame levels.

FIG. **4** is a diagram **90** of control system of a high level example of the operational flow for an approach of changing between three flame excitation voltage levels—high, nominal, and low. The control may typically operate at the nominal voltage level unless the flame drops below a critical threshold, at which time the excitation voltage may adjust to a higher level for greatest accuracy as shown in FIG. **2**. On the other hand, if the flame increases to a higher, less critical level, the excitation voltage may adjust down to a lower level and reduce stress on components. Nominal may be regarded as between low and high.

Flow diagram **90** in FIG. **4** of a control system which may be run by controller **43** of FIG. **1** may begin with a symbol **91** which asks whether the flame is in a critical range. If the answer is yes, then the flame voltage is a high voltage at block **92**, which means the flame scaling is high as indicated in block **93**. Then the system may return to symbol **91** to inquire again whether the flame is in the critical range. If the answer is no, then the system may go to symbol **94** which asks whether the flame is greater than the high flame threshold. If the answer is yes, then the flame voltage is equal to a low voltage as indicated by block **95**, which means that the flame scaling is low as indicated in block **96**. Then the system may return to symbol **91** to inquire again whether the flame is in the critical range. If the answer is no, then the system may go to symbol **94** which asks whether the flame is greater than the high flame threshold. If the answer is no, then the flame voltage is equal to the nominal voltage as indicated by block **97**, which means that the flame scaling is nominal as indicated in block **98**. The system may return to symbol **91** and repeat the inquiries and indications about the flame, voltage and scaling.

FIG. **5** is a diagram of a graphic example of the voltage adjustment of the control system described in diagram **90** of FIG. **4** based on a typical appliance run cycle. The top curve **100** shows the flame current of an appliance as it slowly increases at first through the beginning zone **101**, the critical zone **102** and nominal zone **103**, stabilizes at a high zone **104** level, and then drops off during zones **105** and **106** at the end of the cycle. The control flame voltage is shown on the bottom curve **110** and may be adjusted depending on whether the flame is in the critical, nominal, or high zone or range **102**, **103** or **104**, respectively.

FIG. **6** is a diagram of a graphic example of the control sampling **111** of the flame signal at various times, durations or zones **101**, **102**, **103**, **104**, **105** and **106**, during a typical appliance run cycle. Since the flame signal may be inherently unstable, especially in appliances that have a lot of air movement, it is important to take enough samples to accurately sense the flame. During generally normal running conditions such as in zones **103**, **104** and **105**, the flame just needs to be sampled periodically **111** to maintain normal operation, for example only 20 percent or some of the time, thus reducing stress on the flame components. If the flame has reached a critical level in zone **102** or **106**, the flame sampling **111** may become continuous to ensure the flame is sensed accurately and quickly.

In the present specification, some of the matter may be of a hypothetical or prophetic nature although stated in another manner or tense.

Although the present system has been described with respect to at least one illustrative example, many variations and modifications will become apparent to those skilled in the art upon reading the specification. It is therefore the intention that the appended claims be interpreted as broadly as possible in view of the prior art to include all such variations and modifications.

What is claimed is:

1. A system for optimal flame sensing, comprising:
a flame sensor;
a variable voltage supply connected to the flame sensor; and
a processor connected to the flame sensor and the variable voltage supply; and
wherein:
the flame sensor measures a flame with greater precision with increased voltage applied to the flame sensor; and
the processor determines whether a flame measurement requires greater precision with an increase of voltage provided by the variable voltage supply to the flame sensor.

2. The system of claim **1**, wherein readings of flame sensors of different configurations tend to converge to a same indication as the voltage applied to the sensors increases.

3. The system of claim **1**, wherein the processor proceeds through the steps comprising:
determining whether a flame, if sensed, requires more precise measurement;
if the flame does not require more precise measurement and the flame is not greater than a designated high flame threshold, then the voltage supply changes the voltage applied to the flame sensor toward, to or less than a nominal level;
if the flame requires more precise measurement, then the voltage supply changes the voltage applied to the flame sensor to a higher than nominal level; and
if the flame does not require more precise measurement and the flame is greater than the designated high flame

threshold, then the voltage supply changes the voltage applied to the flame sensor to a lower than nominal level; and
wherein the processor designates the high flame threshold and the nominal level at least in part in accordance with properties of the flame.

4. The system of claim **1**, wherein a flame scaling is determined in accordance with a relationship relative to the voltage applied to the flame sensor.

5. The system of claim **1**, wherein:
data from flame sensor readings at or below a nominal voltage level and a formula provide a basis for calculating equivalent values of the flame sensor as if it were at a voltage higher than the nominal voltage level; and
the processor designates the nominal voltage level at least in pa; by properties of the flame.

6. The system of claim **1**, wherein flame level readings from the flame sensor are from sampled readings for continuous periods of time when more precise measurements are needed, and from sampled readings for shorter, periodic times when more precise measurements are not needed, as determined by the processor.

7. A method for optimal flame sensing, comprising:
taking a first flame reading of a flame at a given level with a flame sensor at a first voltage; and
taking a second flame reading of the flame at the given level with the flame sensor at a second voltage; and
wherein:
the second voltage is greater than the first voltage; and
accuracy of a flame reading is a function of a voltage connected to the flame sensor, the greater the voltage within a certain range, the more accurate is the flame reading.

8. The method of claim **7**, further comprising:
dividing the first flame reading by the first voltage to obtain a first ratio;
dividing the second flame reading by the second voltage to get a second ratio;
dividing the first ratio by the second ratio to obtain a third ratio; and
arranging a relationship for determining a second flame reading from the first flame reading, first voltage, second voltage and third ratio.

9. The method of claim **7**, wherein:

$$r = (R_1/V_1)/(R_2/V_2)$$

R_1 is the first flame reading;
R_2 is the second flame reading;
V_1 is the first voltage;
V_2 is the second voltage;
$V_2 > V_1$; and
$R_{2Scaled} = R_2/r$.

10. The method of claim **9**, further comprising calculating R_2 from one or more other R_1 readings of the flame at one or more other levels and/or one or more other voltages at the flame sensor, respectively.

11. A system for providing flame sensing, comprising:
a flame sensing device for providing measurements of a flame; and
a processor connected to the flame sensing device for receiving measurements of the flame and for controlling voltage at the flame sensing device; and
wherein:
an amount of time that a voltage higher than a nominal voltage is applied to the flame sensing device is minimized; and

the processor determines the nominal voltage at least in part from properties of the flame.

12. The system of claim **11**, further comprising a variable voltage supply, connected to the processor and the flame sensing device, for providing a voltage to the flame sensing device.

13. The system of claim **12**, wherein an increase of voltage to the flame sensing device improves accuracy of measurements of a flame.

14. The system of claim **12**, wherein if accuracy of a flame measurement needs to be increased, then the voltage applied to the flame sensing device is increased.

15. The system of claim **14**, wherein a need for accuracy of a flame measurement increases when the flame decreases.

16. The system of claim **12**, further comprising:

a program executable by the processor; and

wherein the program comprises data and a formula for calculating a measurement of the flame as if a voltage greater than the nominal voltage were applied to the flame sensing device, from a measurement of the flame of the flame sensing device at a voltage equal to or less than the nominal voltage.

17. The system of claim **16**, wherein:

the data and formula comprise:

a first new measurement of a flame at a first voltage; and

a second new measurement of the flame at a second voltage;

$$r = (M_1/V_1)/(M_2/V_2)$$

V_1 is the first voltage;

V_2 is the second voltage;

M_1 is the first new measurement;

M_2 is the second new measurement; and

$M_{2scaled} = M_2/r$.

18. The system of claim **11**, wherein:

the samples of flame current are continuous when accuracy of measurements of a flame is to be higher than a nominal accuracy;

the samples of flame current are periodic when the accuracy of measurements of a flame is to be equal to or less than the nominal accuracy; and

the nominal accuracy is determined by the processor at least in part according to properties of the flame as sensed by the flame sensing device.

19. The system of claim **18**, wherein periodic means that the total samples taken when the flame is present at the flame sensing device is less than the maximum number of samples the processor can handle.

20. The system of claim **18**, wherein periodic means that samples are taken at less than 50 percent of a period of time when the flame is present at the flame sensing device.

* * * * *